

Towards Agrobots: Identification of the Yaw Dynamics and Trajectory Tracking of an Autonomous Tractor

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Abstract

More efficient agricultural machinery is needed as agricultural areas become more limited and energy and labor costs increase. To increase their efficiency, trajectory tracking problem of an autonomous tractor, as an agricultural production machine, has been investigated in this study. As a widely used model-based approach, model predictive control is preferred in this paper to control the yaw dynamics of the tractor which can deal with the constraints on the states and the actuators in a system. The yaw dynamics is identified by using nonlinear least squares frequency domain system identification. The speed is controlled by a proportional-integral-derivative controller and a kinematic trajectory controller is used to calculate the desired speed and the desired yaw rate signals for the subsystems in order to minimize the tracking errors in both the longitudinal and transversal directions. The experimental results show the accuracy and the efficiency of the proposed control scheme in which the euclidean error is below 40 cm for time-based straight line trajectories and 60 cm for time-based curved line trajectories, respectively.

Keywords: Model predictive control, autonomous tractor, agricultural vehicle, agrobots.

1. Introduction

One of the most important tasks in tractor operation is the accurate steering during field operations, e.g. accurate trajectory following during tillage, to avoid damaging the crop or planting when there is no crop yet. Besides, the rows must

16 be parallel, and the distance differences between them must be equal with re-
17 spect to each other during the planting. Moreover, the tractor has to cover the
18 full field without overlap during other operations. However, the steering accu-
19 racy decreases when the operator gets tired or does more actions than driving the
20 tractor like operating/controlling the implements. In order to automate the trajec-
21 tory following problem and also increase the steering accuracy, several automatic
22 guidance systems have been developed to avoid the problems mentioned above.

23 There are various reasons why the control of tractors with a high efficiency is
24 a challenging task. First, an autonomous tractor can be configured with different
25 types of implements and also encounter various environmental conditions (such
26 as humidity, temperature, etc.) during field operations. In such conditions, there
27 is always a trade-off between performance and robustness when a conventional
28 controller, e.g. proportional-integral-derivative (PID) controller, is used. Since
29 conventional controllers have time invariant coefficients and do not have the abil-
30 ity to adapt to changing conditions, they are not appropriate to be used in such
31 agricultural production machines. Second, these machines show many nonlin-
32 ear behaviors such as saturation, dead-time and time lags, which are difficult to
33 handle with conventional control algorithms. Third, tractor navigation involves
34 two subsystems, namely: the yaw dynamics and the longitudinal dynamics which
35 make the control operation more challenging. There is also interaction apart from
36 the hydraulic driveline as a change in the longitudinal speed will change the yaw
37 dynamics and vice versa.

38 In model-based control, the control performance highly depends on the ac-
39 curacy of the model describing the system behavior. In the last decade, several
40 models have been proposed where the yaw dynamics of a wheeled vehicle are
41 described with a bicycle model. Simple kinematic models have been proposed in
42 (OConnor et al., 1996). These models neglect the side-slip of the tires and the
43 dynamics of the steering actuator. Therefore, they are not appropriate for slippery
44 surfaces which are common in field conditions with loose or wet soil. As a so-
45 lution to this problem, a bicycle dynamics model which takes the lateral forces
46 into account is proposed in (OConnor, 1997). As the effect of side slip can be
47 taken into account by this model, it covers a range of slippery and hard surfaces.
48 However, in the previous approach, side slip angles cannot be calculated when the
49 longitudinal speed is equal to zero. As a solution to this problem, the relaxation
50 length approach is proposed in (Karkee and Steward, 2010) to calculate the side-
51 slip angles more accurately. Bevly (Bevly et al., 2002) reported that the relaxation
52 length for only the front tire is adequate in order to model the real-time system.

53 Modeling of side-slip angle, which is the difference between the real and effec-

54 tive steering angle, and determining cornering stiffness values are very important
55 steps in analyzing the yaw dynamics of autonomous vehicles. In (Fang et al.,
56 2011), the cornering stiffness is estimated by a robust adaptive Luenberger ob-
57 server and a sliding mode controller is designed based-on chained system theory.
58 The proposed controller and observer were reported to be robust to time vary-
59 ing lateral disturbances and also inaccurate side-slip angles. As an alternative
60 approach to controlling the agricultural production machines, model reference
61 adaptive control approaches have been proposed in (Derrick and Bevly, 2009).
62 It is observed that the model reference adaptive control algorithm is able to adapt
63 itself to various implementation configurations properly to control lateral position
64 of a tractor for a straight path. In (Gartley and Bevly, 2008), the effect of the
65 hitch point loading on the tractor dynamics is investigated by using a cascaded
66 estimator approach. The experimental results show that the online estimation for
67 the changes in the system provides the ability of adapting the controller gain to
68 maintain the consistent yaw dynamic control of the tractor.

69 Model predictive control (MPC) has been widely used in the chemical process
70 industry since the 1980s. The main goal of MPC approach is to minimize a perfor-
71 mance criterion with respect to constraints of a system’s inputs and outputs. The
72 future values of the system are calculated based on a model. The main advantages
73 of MPC over conventional controllers for the control of agricultural machines are
74 the ability to deal with constraints and with multi-input-multi-output controllers.
75 Several successful applications on agricultural production machines have been re-
76 ported in literature. An MPC design was implemented on the cruise control of
77 a combine harvester (Coen et al., 2008) in which the speed model was devel-
78 oped based on relating the engine speed and the current to the hydraulic pump to
79 the longitudinal speed. The engine speed and the pump settings were controlled
80 simultaneously and this approach was tested experimentally on a New Holland
81 combine harvester. The experimental results show that a satisfactory acceleration
82 performance can be achieved even by keeping the engine speed low. In (Lenain
83 et al., 2005), an MPC strategy is described for the control of an autonomous trac-
84 tor by using an extended kinematic model. This control scheme has been tested
85 experimentally on a farm tractor whose realtime localization is achieved relying
86 solely upon a real-time kinematic (RTK) global positioning system (GPS). How-
87 ever, the control accuracy is limited, because the model used is a kinematic model,
88 and thus neglects the dynamic behaviour of the system. As an extension to MPC,
89 a nonlinear MPC (NMPC) is proposed to obtain better lateral position accuracy
90 of a tractor-trailer system in (Backman et al., 2012). The lateral position error of
91 the trailer was reported to be less than 10 cm in straight paths for a space-based

trajectory in real-time experiments. Moreover, centralized, decentralized and distributed NMPC approaches have respectively been proposed in (Kayacan et al., 2015a, 2014b, 2015b). The drawback of these studies is the same as (Lenain et al., 2005) which is that the model used does not include the dynamic behaviour of the system. Another NMPC algorithm is proposed for the yaw dynamics control of an autonomous vehicle in (Canale et al., 2011). Although it was reported that the proposed controller would allow to use hard constraints for obstacle avoidance strategies, it does not include any real time experiments. Since more advanced control algorithms and mathematical models bring not only more accuracy, but also more computational burden to the real time systems, there always exists a trade-off between the complexity of the method and the computational efficiency of the overall system.

The main contributions of this study beyond the state of the art are modeling the yaw dynamics of an autonomous tractor considering various definitions of side slip angles and controlling it with good computational efficiency. In order to achieve this, first, the yaw dynamics model of the autonomous tractor has been derived, the model structures have been validated, and model parameters have been estimated by using frequency response function (FRF) measurements. Finally, the nonlinear least square (NLS) frequency domain identification (FDI) approach is used to obtain the model parameters to determine which model is better for the tractor at hand. After the identification of the yaw dynamics, an MPC controller for the yaw dynamics is designed based on the identified model. Then, this yaw dynamics controller has been combined with a kinematic controller for the trajectory tracking in which the kinematic controller is used for compensating the errors both in the x- and y-axes.

This paper is organized as follows: The experimental set-up is described in Section II. The kinematic model of the system and the mathematical model of the yaw dynamics are presented in Section III. In Section IV, the identification of the yaw dynamics is described. In Section V, the basics of the implemented MPC approach are given. The overall control structure and the real-time experimental results are presented in Section VI. Finally, some conclusions are drawn from this study in Section VII.

2. Experimental Set-up Description

The aim of this study is to track a time-based trajectory with a small agricultural tractor shown in Fig. 1. The GPS antenna is located straight up the center of the tractor rear axle to provide highly accurate position information for the



Figure 1: The experimental set-up (CNH TZ25DA)

128 autonomous tractor. The height of the antenna is 2 m above ground level. It is
 129 connected to a Septentrio AsteRx2eH RTK-DGPS receiver (Septentrio Satellite
 130 Navigation NV, Leuven, Belgium) with a specified position accuracy of 2 cm at
 131 a 20-Hz sampling frequency. The Flepos network supplies the RTK correction
 132 signals via internet by using a *Digi Connect WAN 3G* modem.

133 The block diagram of the hardware is shown in Fig. 2. The GPS receiver and
 134 the internet modem are connected to a real time operating system (PXI platform,
 135 National Instruments Corporation, Austin, TX, USA) through an RS232 serial
 136 communication. The PXI system acquires the steering angle, the GPS data and
 137 controls the tractor by sending messages to actuators. A laptop connected to the
 138 PXI system by WiFi functions as the user interface of the autonomous tractor. The
 139 algorithms are implemented in *LabVIEWTM* version 2011 (National Instruments,
 140 Austin, TX, USA). They are executed in real time on the PXI and updated at a
 141 rate of 20-Hz.

142 The designed MPC in Section 5 calculates the desired steering angle for the
 143 front wheels, and a low level controller, a PI controller in our case, is used to
 144 control the steering mechanism. In the inner closed loop, the steering mecha-
 145 nism is controlled by using an electro-hydraulic valve with a maximal flow of 12
 146 liter/min. The electro-hydraulic valve characteristics are highly nonlinear and in-
 147 clude a saturation and a dead-band region. The voltage limited between 0 – 12
 148 volt and the steering angle limited between $\pm 45^\circ$ became the input and the output

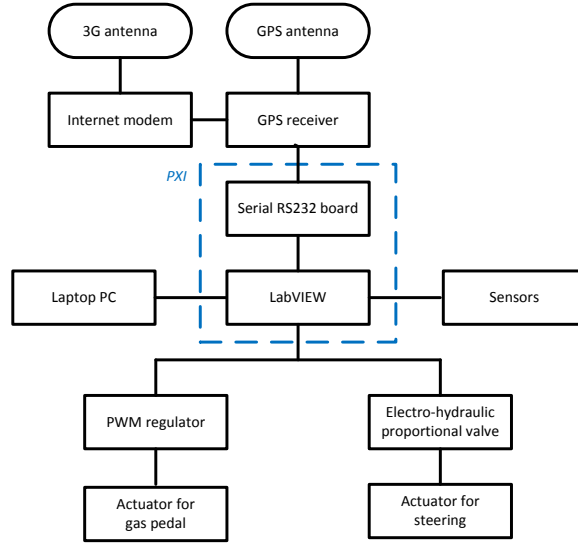


Figure 2: Block diagram of hardware

149 for the steering system, respectively. The angle of the front wheels is measured
150 using a potentiometer mounted on the front axle, yielding an angle measurement
151 resolution of 1° .

152 The speed of the tractor is controlled by using an electro-mechanic valve.
153 There are two PID type controllers in speed control. The PID controllers in the
154 outer closed-loop and the inner closed-loop are generating the desired pedal posi-
155 tion with respect to the speed of the tractor and voltage for the electro-mechanic
156 valve with respect to the pedal position, respectively. Since the measured speed
157 coming from the GPS is noisy, a discrete Kalman Filter (KF) was used to re-
158 duce noise. A position-velocity model described in (Brown and Hwang, 2012)
159 was used where vehicle velocity is assumed as a random-walk process. The KF
160 assumes that the vehicle moves with a constant velocity between discrete-time
161 steps. The state vector of the model used in the KF and the state transition matrix
162 are written as follows:

$$\begin{aligned}
\hat{\mathbf{x}}_{k+1} &= \Phi(T_s)\hat{\mathbf{x}}_k \\
&= \begin{bmatrix} 1 & T_s & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & T_s \\ 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_k \\ v_{x,k} \\ y_k \\ v_{y,k} \end{bmatrix}
\end{aligned} \tag{1}$$

where $\Phi(T_s)$, $v_{x,k}$ and $v_{y,k}$ are the state transition matrix and velocities coming from the GPS, respectively.

3. Mathematical Model of an Autonomous Tractor

3.1. Kinematic Model

The schematic diagram of an autonomous tractor is illustrated in Fig. 3. The linear velocities \dot{x} and \dot{y} at the rear axle of the tractor (point R) are written as follows:

$$\begin{aligned}
\dot{x}_R &= v_x \cos \psi \\
\dot{y}_R &= v_x \sin \psi \\
\dot{\psi}_R &= \frac{v_x \tan \delta}{L}
\end{aligned} \tag{2}$$

where v_x , ψ , δ and L represent the longitudinal velocity, the yaw angle of the tractor, the steering angle of the front wheel, the distance between the front axle and the rear axle of the tractor, respectively.

Instead of considering the point R in the dynamic equations and the trajectory tracking control, the center of gravity (CG) is preferred. Thus, the linear velocities \dot{x} and \dot{y} are projected onto the CG as follows:

$$\begin{aligned}
\dot{x} &= v_x \cos \psi - v_y \sin \psi \\
\dot{y} &= v_x \sin \psi + v_y \cos \psi
\end{aligned} \tag{3}$$

where v_y is the lateral velocity of the tractor at the CG.

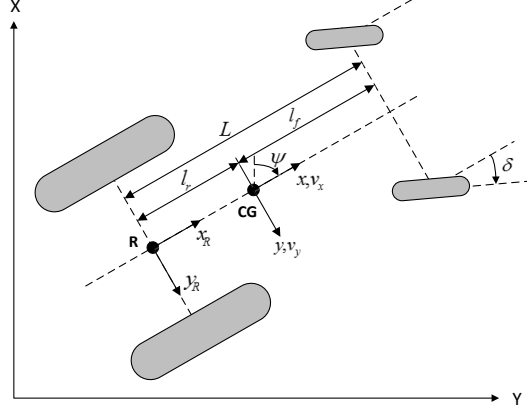


Figure 3: Kinematic model for an autonomous tractor

3.2. Modeling of the Yaw Dynamics

As the driving speed of the tractor is rather limited, it is reasonable to assume that the lateral forces on the right and left wheels are equal to each other and can be summed. Therefore, the tractor is modelled in 2D as a bicycle system. The velocities, the sideslip angles and the forces on the rigid body of an autonomous tractor are schematically illustrated in Fig. 4. The yaw dynamics models are derived based on the following assumptions:

- The traction forces are neglected,
- The aerodynamic forces are neglected,
- The tire moments are small, such that these can be neglected,
- The pitch and roll dynamics are neglected.

The notations used in the following (see also Fig. 4) are summarized in Table 1.

3.2.1. Vehicle Dynamics

The lateral motion of the tractor is written as follows:

$$m(\dot{v}_y + v_x \gamma) = F_{t,f} \sin \delta + F_{l,f} \cos \delta + F_{l,r} \quad (4)$$

Table 1: NOMENCLATURE

m	Mass
I	Moment of inertia of the tractor around the vertical axis
v_x	Longitudinal velocity of the CG
v_y	Lateral velocity of the CG
ψ	Yaw angle of the tractor
γ	Yaw rate of the tractor
δ	Steering angle of the front wheels
l_f	Distance between the front axle and the CG
l_r	Distance between the rear axle and the CG
L	Distance between the front axle and the rear axle
$F_{t,f}$	Traction force on the front wheels
$F_{t,r}$	Traction force on the rear wheels
$F_{l,f}$	Lateral force on the front wheels
$F_{l,r}$	Lateral force on the the rear wheels
$C_{\alpha,f}$	Cornering stiffness of the front wheels
$C_{\alpha,r}$	Cornering stiffness of the rear wheels
α_f	Side slip angle of the front wheels
α_r	Side slip angle of the rear wheels
σ_f	Relaxation length of the front wheels
σ_r	Relaxation length of the rear wheels

192 where γ , m , $F_{t,f}$, $F_{l,f}$, $F_{l,r}$ represent the yaw rate and the mass of the tractor, the
193 traction and lateral forces on the front wheel, and the lateral force on the rear

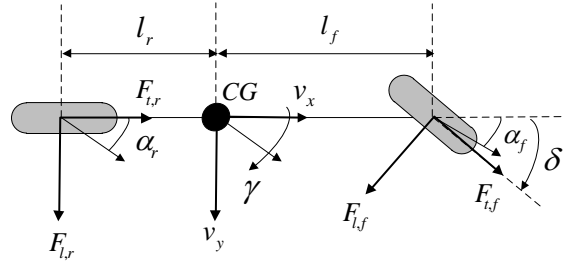


Figure 4: Bicycle dynamics model for a tractor system: velocities, side slip angles and forces on the rigid body of the system

194 wheel.

195 The yaw motion of the tractor is written as follows:

$$I\dot{\gamma} = l_f(F_{t,f} \sin \delta + F_{l,f} \cos \delta) - l_r F_{l,r} \quad (5)$$

196 where l_f , l_r and I represent the distance between the front axle and the CG of
 197 the tractor, and the distance between the rear axle and the CG of the tractor, the
 198 moment of inertia of the tractor. Since it is a time consuming process to calculate
 199 the inertial moment of the tractor, an approximate value for the inertial moment
 200 can be calculated as follows (Garrott et al., 1998):

$$I = ml_f l_r \quad (6)$$

201 3.2.2. Tire Model

202 The lateral tire forces are calculated in a linear model in which they are as-
 203 sumed to be proportional to the slip angles (Piyabongkarn et al., 2009; Geng et al.,
 204 2009) :

$$F_{l,i} = -C_{\alpha,i} \alpha_i \quad i = \{f, r\} \quad (7)$$

205 where $C_{\alpha,i}$ and α_i $i = \{f, r\}$, represent the cornering stiffnesses of the tires and
 206 the side-slip angles, respectively.

207 The tire side-slip angles must be calculated in order to determine the slip
 208 forces. The side slip angles of the front and rear tires have been considered in
 209 a linear form, and they are written as follows:

$$\alpha_f = \frac{v_y + l_f \gamma}{v_x} - \delta \quad (8)$$

$$\alpha_r = \frac{v_y - l_r \gamma}{v_x} \quad (9)$$

210 As can be seen from the equations above, the side slip angles cannot be calculated
 211 when the longitudinal speed is zero. As a solution to this problem, the relax-
 212 ation length is defined as the amount a tire rolls to reach the steady state side slip
 213 angle. Previous research in vehicles suggests that the relaxation length of a tire
 214 plays a very important role in the steering motion at high velocities (Owen and
 215 Bernard, 1982; Crolla, 1983). This result motivates us to use this approach for
 216 the agricultural vehicles due to the fact that the calculation of the side-slip angles

217 in (8) and (9) for very low velocities can go to infinity. Since a tire generates
 218 the steady state side slip angle simultaneously, a first order mathematical model
 219 is used to describe the slip angle dynamics through the relaxation length. A first
 220 order differential equation for the side slip angle can be written as follows:

$$\dot{\alpha} = \frac{v_x}{\sigma}(\alpha_0 - \alpha) \quad (10)$$

221 A relaxation length of 1.5 times the tire radius has been proposed for agri-
 222 cultural vehicles as it allows to obtain similar changes for a similar increase in
 223 velocity (Bevly et al., 2002). For passenger vehicles which have higher velocity
 224 than agricultural vehicles, a factor larger than 2 is typically selected (Loeb et al.,
 225 1990).

226 By combining equations (8), (9) and (10), the time derivatives of the side slip
 227 angles of the front and rear wheels can be written as follows:

$$\dot{\alpha}_f = \frac{v_y + l_f \gamma - v_x(\delta + \alpha_f)}{\sigma_f} \quad (11)$$

$$\dot{\alpha}_r = \frac{v_y - l_r \gamma - v_x \alpha_r}{\sigma_r} \quad (12)$$

228 where σ_f and σ_r represent the relaxation length of the front and rear tires of the
 229 tractor, respectively.

230 3.2.3. Equations of Yaw Motion

231 The tractor is driven on the field with a constant longitudinal velocity as gen-
 232 erally required in automatic guidance of agricultural vehicles. For this reason, the
 233 time derivative of the longitudinal velocity \dot{v}_x can be set to zero and the longitu-
 234 dinal velocity v_x can be assumed as a parameter. It is also assumed that the steering
 235 angle is sufficiently small to justify linearization of the equations. Thus, (4) and
 236 (5) can be written considering (7) as follows:

$$\begin{aligned} m\dot{v}_y &= -mv_x\gamma - C_{\alpha,f}\alpha_f - C_{\alpha,r}\alpha_r \\ I\dot{\gamma} &= -l_f C_{\alpha,f}\alpha_f + l_r C_{\alpha,r}\alpha_r \end{aligned} \quad (13)$$

237 By using different considerations, three different transfer functions can be
 238 written. Firstly, the traditional bicycle model for the yaw motion of the au-
 239 tonomous tractor can be obtained by combining (8), (9) and (13). The relation

240 between the yaw rate of the autonomous tractor and the steering angle of the front
 241 wheels can be written in transfer function form as follows:

$$G_{TB}(s) = \frac{b_1^*s + b_0^*}{a_2^*s^2 + a_1^*s + a_0^*} \quad (14)$$

242 Secondly, the bicycle model with relaxation length approach for only the front
 243 wheel can be obtained by combining (9), (11) and (13). The relation between the
 244 yaw rate of the autonomous tractor and the steering angle of the front wheels can
 245 be written in transfer function form as follows:

$$G_{RLF}(s) = \frac{b_1^\diamond s + b_0^\diamond}{a_3^\diamond s^3 + a_2^\diamond s^2 + a_1^\diamond s + a_0^\diamond} \quad (15)$$

246 Thirdly, the bicycle model with the relaxation length approach for the front and
 247 the rear wheels can be obtained by combining (11), (12) and (13). The relation
 248 between the yaw rate of the autonomous tractor and the steering angle of the front
 249 wheels can be written in transfer function form as follows:

$$G_{RLFR}(s) = \frac{b_2^*s^2 + b_1^*s + b_0^*}{a_4^*s^4 + a_3^*s^3 + a_2^*s^2 + a_1^*s + a_0^*} \quad (16)$$

250 The relation between the physical parameters in the previous equations and
 251 the transfer function parameters in (15) and (16) is shown in Appendix A. These
 252 different models might be proper for different real-time systems or different cases
 253 of the same real-time system. In Section IV these transfer functions (14), (15)
 254 and (16) will be fit to the empirical transfer function estimate obtained from the
 255 frequency domain experiments to decide which is most appropriate for the real-
 256 time system considered in this study.

257 4. Identification of the Yaw Dynamics

258 It was observed during the identification experiments that the front wheels
 259 reached their limits due to drift when the excitation signal was applied to the steer-
 260 ing mechanism in an open-loop fashion. As a solution to this drifting problem, the
 261 system was controlled with a P controller, and then the closed-loop system was
 262 identified. The schematic diagram of the identification processes is shown in Fig.
 263 5. The steering mechanism of the tractor was identified in (Kayacan et al., 2014c,
 264 2013).

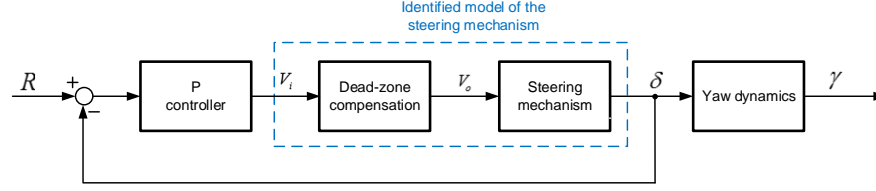


Figure 5: The schematic diagram of the identification processes

265 In this section, the nonlinearity of the yaw dynamics is checked by using an
 266 odd-odd multisine signal and a frequency content in which the linear contribu-
 267 tions are dominant is determined for the identification process (Vanhoenacker and
 268 Schoukens, 2003; Schoukens et al., 2005). A multisine signal is applied to the
 269 steering angle controller, and the linear models with respect to Section 3.2.3 are
 270 obtained by using the NLS FDI method (Kayacan et al., 2013). After the fre-
 271 quency domain analysis, a qualitative comparison of the linear models in time-
 272 domain is given, and the effect of the longitudinal speed is shown.

273 After the design and the implementation of the steering angle controller, the
 274 frequency spectrum of the response of the yaw rate to a random odd-odd mul-
 275 tisine excitation is shown in Fig. 6. It can be seen that the contribution of the
 276 nonlinearities to the total response is as large as the linear contribution after 2 Hz.
 277 Moreover, it is known that the noise in FRF measurements above 1.5 Hz will be
 278 due to the lack of input signal. As can be seen from Fig. 7, the range till 2 Hz
 279 is not enough to capture the peak, and the magnitude is still quite high. Since the
 280 nonlinear contributions are dominant after 2 Hz, a linear model can be derived
 281 until 2 Hz.

282 Based-on the considerations above, a multisine signal with a frequency content
 283 between 0.02Hz and 2Hz has been applied to the system as an excitation signal.
 284 The model parameters are identified by using the NLS FDI approach based on
 285 FRF measurements. In Fig. 7, the measured FRF and the FRFs of identified
 286 models are shown. As can be seen from Fig. 7, since the traditional bicycle model
 287 in (14) consists of one zero and two poles such a system cannot have a frequency
 288 response like the one in Fig. 7, and thus it is not a candidate for the parameter
 289 estimation. The other two models in (15) and (16) can be proper candidates for
 290 the parameter estimation.

291 After the identification process, the transfer functions of the identified yaw

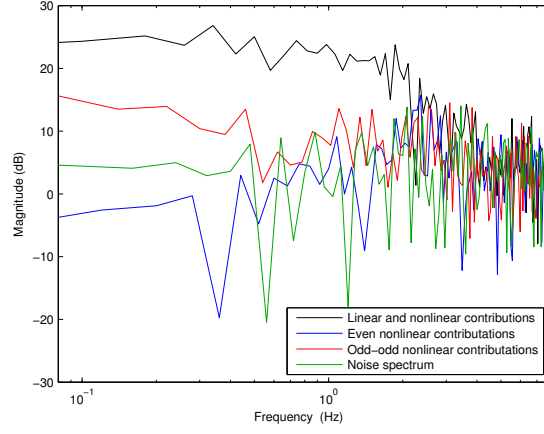


Figure 6: The analysis of nonlinear contributions

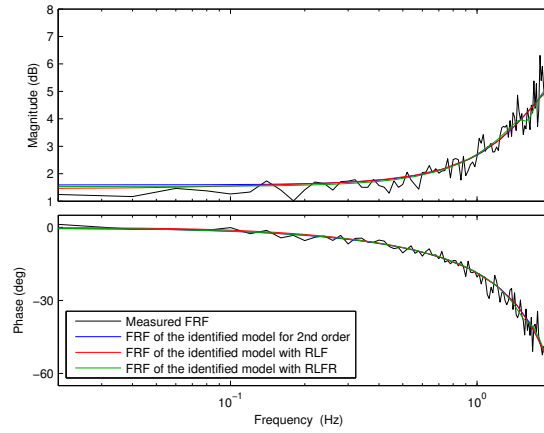


Figure 7: Measured FRF and FRF of the identified models

292 dynamics model are calculated respectively for the bicycle model with relaxation
 293 length approach for only the front wheel, the bicycle model with relaxation length
 294 approach for the front and the rear wheels, and a second order system as follows:

$$G_{RLF}(s) = \frac{292s + 177}{s^3 + 11.6s^2 + 249s + 150} \quad (17)$$

$$G_{RLFR}(s) = \frac{279s^2 + 335s + 27860}{s^4 + 11.5s^3 + 347s^2 + 1311s + 23340} \quad (18)$$

$$G_{2nd}(s) = \frac{291}{s^2 + 10.9s + 242} \quad (19)$$

Even if the models derived in (15) and in (16) fit with the real-time FRF measurements, these models are higher order transfer functions. As a simplified model, a second order transfer function in (19) is fitted to FRF measurements. As can be seen from Fig. 7, the frequency domain response of the proposed transfer function in (19) is similar to the derived models in (17) and (18). It can be concluded that a second order transfer function can also be used to model the yaw dynamics of the tractor at hand. It should not go unnoticed that we do not throw the physical models away by proposing the second order empirical transfer function. The reason is that the empirical model might not be suitable for different soil conditions due to the higher values for the slip. For a better understanding, (11), (12) and (13) should be considered carefully in which there are four ordinary differential equations that describe the yaw dynamics behaviour of the tractor. In case of having small side-slip angles, the equations for the side-slips can be neglected. As a result, there will only be two ordinary differential equations for the yaw dynamics resulting in a second order transfer function. Another disadvantage of using the empirical model is that the side-slip angles cannot be estimated.

A qualitative comparison of the three identified models in time-domain is given in Table 2. A multisine signal between 0 and 8 Hz is given to the system to check whether the models are appropriate for the real-time system at hand. As can be seen from Table 2, although the parameter estimation of the three models is done until 2Hz in frequency domain, the models still give reasonable results at higher frequencies. Moreover, it can be concluded that the addition of zeros and poles do not increase the accuracy of the identified models when the side-slips are negligible.

The selected and estimated parameters for the bicycle model with the relaxation length approach for the front and the rear wheels are given in Table 3. It is assumed that the values of the mass m , the distance between the front axle and the CG l_f , and the distance between the rear axle and the CG l_r are known. The inertia moment I is calculated based-on (6). The values for the cornering stiffness and the relaxation length of the front and the rear wheels are approximately estimated

Table 2: Root Mean Square Error of the Identified Models

Model	Multisine Range	RMSE
$G_{2nd}(s)$	0-2Hz	0.0787
	0-8Hz	0.0854
$G_{RLF}(s)$	0-2Hz	0.0787
	0-8Hz	0.0854
$G_{RLFR}(s)$	0-2Hz	0.0793
	0-8Hz	0.0860

327 based-on (18). On the other hand, since it is observed that the parameters are not
 328 realistic for the bicycle model with the relaxation length approach for the front
 329 wheel, they have not been able to be estimated. During the parameter estimation
 330 process, it has been observed that there are more than one solution for a specific
 331 parameter. When two separate solutions resulted in roughly the same values, the
 332 estimated parameters values have been said to be realistic.

Table 3: Numerical Values of the Parameters

Parameter	Unit	Value
m	kg	700
I	$kg\ m^2$	280
l_f	m	1
l_r	m	0.4
$C_{\alpha,f}$	N/rad	8000 ± 500
$C_{\alpha,r}$	N/rad	90000 ± 7000
σ_f	m	0.1942
σ_r	m	1.6657

333 4.1. The effect of longitudinal velocity

334 In order to be able to analyze the effect of the linear longitudinal velocity
 335 on the identified linear models, an additional experiment has been performed in
 336 which the linear velocity is varied from 1 m/s to 2 m/s. In Figure 8, it is illustrated
 337 how the poles and the zeroes of the three models mentioned above change with
 338 respect to the longitudinal velocity. As can be seen from Fig. 8, the bicycle model
 339 with RLFR and the bicycle model with RLF have pole-zero cancelation, and the

poles go to the left on the s-plane, which is an expected case from the given transfer functions in (15) and (16), when the longitudinal velocity increases.

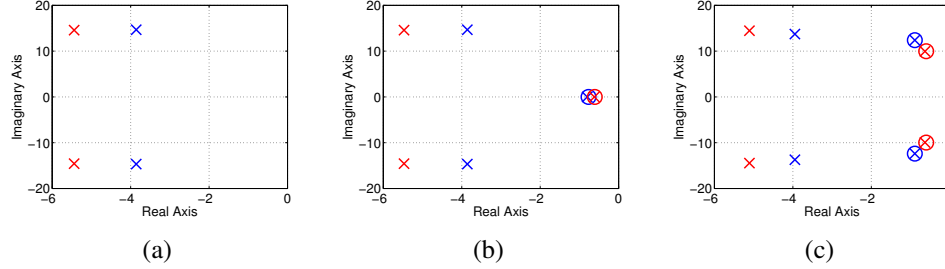


Figure 8: Poles and zeroes at 1 m/s (blue) and 2 m/s (red): (a) 2nd order model (b) the bicycle model with RLF (c) the bicycle model with RLFR

5. Model Predictive Control

MPC controllers predict the future system behavior based-on the system model, and calculate the optimal input sequence based on these predictions (Maciejowski, 2002). As it was stated in the previous section, among the identified models, the second-order empirical model gives similar performance accuracy with the other mentioned physical models to represent the autonomous tractor at hand. Thus, the second-order empirical model is used for the MPC design in this study. The objective function consists of a function of the states, the outputs and the inputs of the system. The control action is calculated by minimizing the cost function subject to the predicted behavior of the model and the system constraints. The states and the outputs are predicted over a given prediction horizon. The main equality constraint is the system model, and the inequality constraints are the state constraints, the output and the input constraints (actuators limits). In our case, there are no constraints on the system states, but the constraint on the input to the system, which is the steering angle, is defined in (21).

A discrete-time linear invariant state-space model can be written as follows:

$$\begin{aligned} x(k+1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned} \quad (20)$$

where $x(k) \in \mathbb{R}^n$, $y(k) \in \mathbb{R}^p$ and $u(k) \in \mathbb{R}^m$ are the state, output and input variables, respectively. The pre-known matrices A , B , C and D are calculated considering the sampling time of the real-time system by using (19).

361 The constraints are written for all $k \geq 0$ as follows:

$$\begin{aligned} -45 \text{ degrees} &\leq u(k) \leq 45 \text{ degrees} \\ -55 \text{ degrees/s} &\leq \Delta u(k) \leq 55 \text{ degrees/s} \end{aligned} \quad (21)$$

362 The cost function in its general form is written as follows:

$$J(U, x(k)) = \sum_{i=0}^{N_p} x_{k+i|k}^T Q x_{k+i|k} + \sum_{i=0}^{N_c-1} u_{k+i}^T R u_{k+i} \quad (22)$$

363 where $N_p = 8$ and $N_c = 3$ represent the prediction and control horizons, and
 364 $U = [u_k^T, \dots, u_{k+N_c-1}^T]^T$ is the vector of the input steps from sampling instant k
 365 to sampling instant $k + N_c - 1$. It was reported that prediction and control hori-
 366 zons are related to the speed of vehicles for a stable performance (Keviczky et al.,
 367 2006). Since a tractor is a slow vehicle, small prediction and control horizons are
 368 chosen to decrease computational burden in real-time applications.

369 The first sample of U is applied to the plant:

$$u^* = u_k^T \quad (23)$$

370 and the optimization problem is solved over a shifted horizon for the next sam-
 371 pling time. Q and R are positive-definite weighting matrices defined as follows:

$$Q = \text{diag}(0.5) \quad , \quad R = \text{diag}(1) \quad (24)$$

372 The following plant objective function is solved at each sampling time for the
 373 MPC:

$$\begin{aligned} \min_{x(\cdot), u(\cdot)} \quad & \sum_{i=0}^{N_p} x_{k+i|k}^T Q x_{k+i|k} + \sum_{i=0}^{N_c-1} u_{k+i}^T R u_{k+i} \\ \text{subject to} \quad & x(k+1) = Ax(k) + Bu(k) \\ & y(k) = Cx(k) + Du(k) \\ & -45 \text{ degrees} \leq \delta(k) \leq 45 \text{ degrees} \\ & -55 \text{ degrees/s} \leq \Delta \delta(k) \leq 55 \text{ degrees/s} \end{aligned} \quad (25)$$

374 In our case, the designed MPC minimizes the error between the reference yaw
 375 rate and the measured yaw rate, and finds the desired steering angle $\delta_{desired}$ to the
 376 real-time system.

377 6. Trajectory Tracking Control

378 6.1. Overall Control Scheme

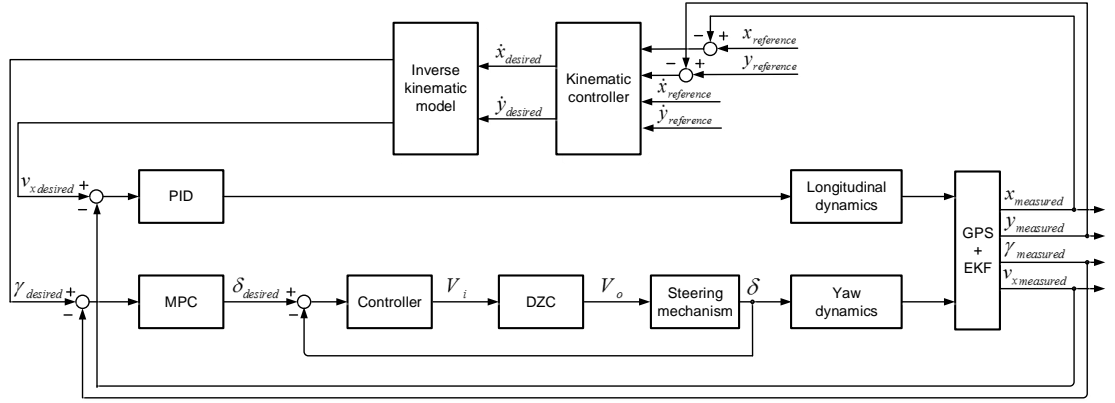


Figure 9: Block diagram of the proposed control scheme

379 6.1.1. Kinematic Controller

380 The kinematic model in (3) is re-written in the algebraic model as follows:

$$\begin{bmatrix} \dot{x} \\ \dot{y} \end{bmatrix} = \begin{bmatrix} \cos \psi & -l_r \sin \psi \\ \sin \psi & l_r \cos \psi \end{bmatrix} \begin{bmatrix} v_x \\ \gamma \end{bmatrix} \quad (26)$$

381 where the lateral velocity v_y equals to γl_r .

382 An inverse kinematic model is needed to generate the desired longitudinal
383 speed and the desired yaw rate for the tractor. The inverse kinematic model in this
384 study is written as follows:

$$\begin{bmatrix} v_x \\ \gamma \end{bmatrix} = \begin{bmatrix} \cos \psi & \sin \psi \\ -\frac{1}{l_r} \sin \psi & \frac{1}{l_r} \cos \psi \end{bmatrix} \begin{bmatrix} \dot{x} \\ \dot{y} \end{bmatrix} \quad (27)$$

385 Considering $\dot{x}_d = \dot{x}_r + k_s \tanh(k_c e_x)$ and $\dot{y}_d = \dot{y}_r + k_s \tanh(k_c e_y)$, the kinematic
386 control law proposed in (Martins et al., 2008; Kayacan et al., 2014a) to be applied
387 to the tractor for the trajectory tracking control is written as:

$$\begin{bmatrix} v_{x_d} \\ \gamma_d \end{bmatrix} = \begin{bmatrix} \cos \psi & \sin \psi \\ -\frac{1}{l_r} \sin \psi & \frac{1}{l_r} \cos \psi \end{bmatrix} \begin{bmatrix} \dot{x}_d \\ \dot{y}_d \end{bmatrix} \quad (28)$$

where v_{x_d} and γ_d are respectively the desired speed and the desired yaw rate, and $e_x = x_r - x$ and $e_y = y_r - y$ are the current position errors in the X – and Y – axes, respectively. The parameters k_c and k_s are the gain and the saturation constant of the kinematic controller, respectively. The advantage of the kinematic model used in this paper is that since it has a saturation, the generated yaw rate cannot have extremely large values. The coordinates (x, y) and (x_r, y_r) are the current and the reference coordinates at the CG of the tractor, respectively.

Considering a perfect velocity tracking ($v_x = v_{x_d}$ and $\gamma = \gamma_d$) which means that the dynamics effects are ignored, the stability analysis is done by using a Lyapunov function (Martins et al., 2008).

6.1.2. Dynamic Controllers

The proposed control scheme used in this study is schematically illustrated in Fig. 9. A PID controller is used for the longitudinal velocity control. In the yaw dynamics control, an MPC controller is designed and its output is the desired steering angle for the front wheels. A low level PI controller is used to control the steering mechanism.

6.1.3. State Estimation

An Extended Kalman Filter (EKF) was used for the state estimation. Since only one GPS antenna was mounted on the tractor, the yaw angle of the tractor cannot be measured. It is to be noted that the yaw angle of the tractor plays a very important role in the accuracy of trajectory tracking control as the estimated yaw angle is used in the inverse kinematic model to generate the desired speed and the desired yaw rate for the system. Therefore, the inputs of the EKF are position and velocity values from GPS. The outputs of the EKF are the position of the tractor on x - and y -coordinate system and the yaw angle. In trajectory control, the estimated values are used. Since the GPS antenna was located at point R on the tractor, the kinematic model in (2) is used. The discrete-time kinematic model used by EKF is written with a sampling interval T_s as follows:

$$\begin{aligned}
x_{k+1} &= x_k + T_s v_{x_k} \cos \psi_k \\
y_{k+1} &= y_k + T_s v_{x_k} \sin \psi_k \\
\psi_{k+1} &= \psi_k + T_s v_{x_k} \frac{\tan \delta_k}{L}
\end{aligned} \tag{29}$$

416 The general form of the estimated system model is written:

$$\begin{aligned}
\hat{x}_{k+1} &= f(\hat{x}_k, u_k) + w_k \\
\hat{y}_{k+1} &= h(\hat{x}_k) + v_k
\end{aligned} \tag{30}$$

417 where f is the estimation model for the system and h is the measurement function.
418 The difference between the kinematic model and estimation model is the process
419 noise w_k and observation noise v_k both in the state and the measurement equations.
420 They are both assumed to be independent and zero mean multivariate Gaussian
421 noises with covariance matrices Q_k and R_k , respectively:

$$\begin{aligned}
w_k &\sim N(0, Q_k) \\
v_k &\sim N(0, R_k)
\end{aligned} \tag{31}$$

422 6.2. Experimental Results

423 An 8-shaped trajectory with both straight and curved line geometries has been
424 applied as the reference trajectory. The motivation of choosing an 8-shaped tra-
425 jectory is that we can evaluate the performance of the controller both for straight
426 and curved lines. The reference and actual trajectories of the autonomous tractor
427 and the error values on the related trajectory are shown in Figs. 10 and 11, respec-
428 tively. The experimental results show that the proposed control scheme is able to
429 control the system.

430 Since the given trajectory consists of linear and curvilinear lines, the optimal
431 values of the Q and R weighting matrices of the MPC controller had to be tuned
432 by making a trade-off between optimal performance on the straight and curvilinear
433 lines. It was observed during the experiments that although the curvilinear
434 trajectory tracking is better with aggressive controllers (having big gains) than the
435 linear trajectory, these controllers give a large overshoot for the linear trajectory.
436 Meanwhile, although the linear trajectory tracking is worse with an aggressive

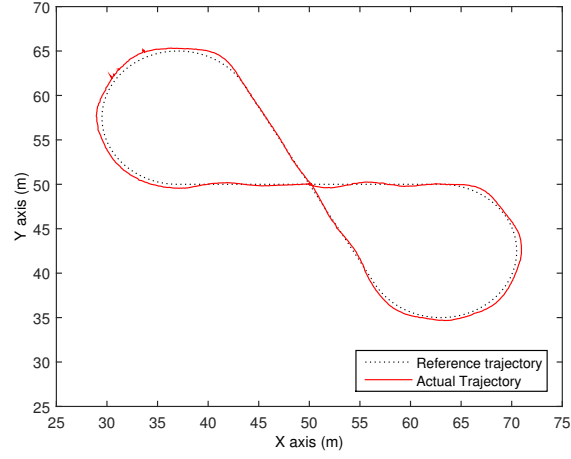


Figure 10: The reference and the actual trajectories of the autonomous tractor

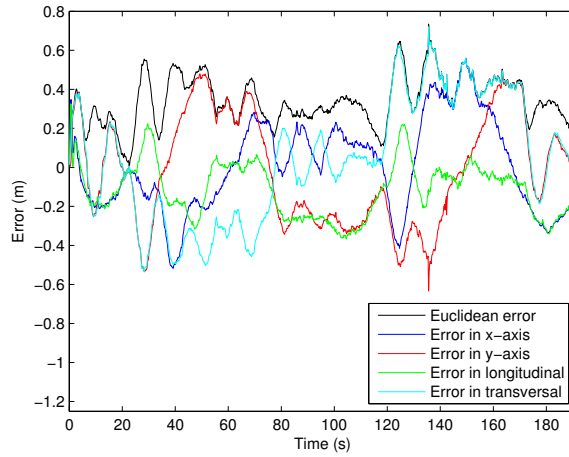


Figure 11: The trajectory following error both in x- and y-axes

437 controller than the curvilinear trajectory, this controller gives large error values
 438 for the curvilinear trajectory. Thus, it can be concluded that if a reference sig-
 439 nal consisting of only linear or curvilinear geometry is considered, more accurate
 440 results can be obtained.

441 Figures 12 and 13 show the longitudinal velocity and the yaw rate responses
 442 of the autonomous tractor. As can be seen from Fig. 12, there is a steady state
 443 error for the control of the longitudinal velocity by the PID controller. On the

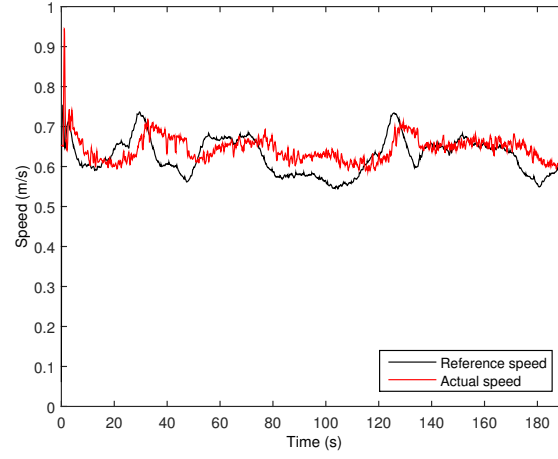


Figure 12: The reference and the actual longitudinal speed of the autonomous tractor

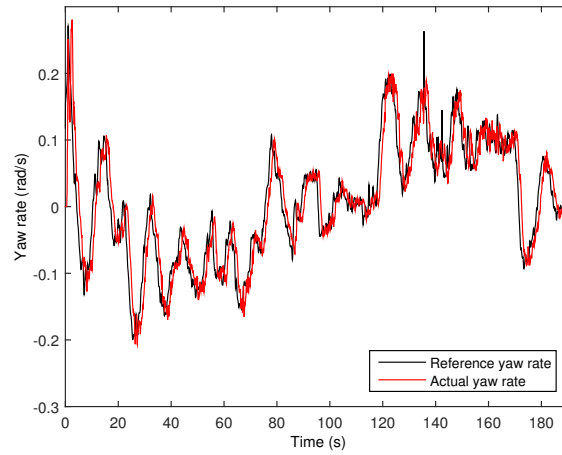


Figure 13: The reference and the actual yaw rate of the autonomous tractor

444 other hand, there is no steady state error for the yaw rate control by the MPC
 445 controller. These figures show that the error in both the x- and y-axes come from
 446 the poor control performance of the PID controller in the longitudinal dynamics,
 447 because it cannot cope with the strong high nonlinearities and interaction between
 448 the subsystems.

449 7. Conclusion

450 In this study, modeling, identification and control aspects of an autonomous
451 tractor have been investigated. Three yaw dynamics models have been derived
452 from the equations of motion of the system. The parameters of these transfer
453 functions have been estimated through nonlinear least squares frequency domain
454 identification. While the fourth order transfer function model derived from the
455 equations of motion considering the cornering stiffness and relaxation length for
456 both the front and rear wheels gave realistic parameters estimates, the transfer
457 function could be reduced to a second order model as not all yaw dynamics could
458 be excited by the steering mechanism. This reduced order transfer function was
459 then incorporated in an MPC for the yaw dynamics control and combined with
460 a PID controller for the longitudinal speed control and an inverse kinematic con-
461 troller for the trajectory tracking. The performance of these controllers was eval-
462 uated during real-time tests. The yaw rate control by the MPC gave satisfactory
463 results, while the PID control of the longitudinal velocity did not. Although the
464 8-shaped time based reference trajectory could be tracked reasonably well, the
465 longitudinal speed control should be improved to obtain better trajectory tracking.
466 In order to increase the control accuracy in the straight lines, a space-based tra-
467 jectory in which the longitudinal speed is constant, but only the yaw rate of the
468 tractor is controlled, can be preferred. The MPC presented in this study provides
469 the ideal framework for this.

470 **Appendix A. The parameters in the transfer functions**

471 The parameters in (15) and (16) are written as follows:

$$\begin{aligned}
b_0^\diamond &= \frac{C_{\alpha,f}C_{\alpha,r}(l_f + l_r)}{Im\sigma_f} \\
b_1^\diamond &= \frac{C_{\alpha,f}l_f v_x}{I\sigma_f} \\
a_0^\diamond &= \frac{mv_x^2(-C_{\alpha,f}l_f + C_{\alpha,r}l_r) + C_{\alpha,f}C_{\alpha,r}(l_f + l_r)^2}{Imv_x\sigma_f} \\
a_1^\diamond &= \frac{I(C_{\alpha,f} + C_{\alpha,r}) + m(C_{\alpha,f}l_f^2 + C_{\alpha,r}l_r^2) + C_{\alpha,r}l_r m\sigma_f}{Im\sigma_f} \\
a_2^\diamond &= Imv_x^2 + C_{\alpha,r}l_r^2 m\sigma_f + IC_r\sigma_f \\
a_3^\diamond &= 1 \\
b_0^* &= \frac{C_{\alpha,f}C_{\alpha,r}(l_f + l_r)v_x}{Im\sigma_f\sigma_r} \\
b_1^* &= \frac{C_{\alpha,f}l_f v_x^2}{I\sigma_f\sigma_r} \\
b_2^* &= \frac{C_{\alpha,f}l_f v_x}{I\sigma_f} \\
a_0^* &= \frac{C_{\alpha,f}C_{\alpha,r}(l_f + l_r)^2}{Im\sigma_f\sigma_r} \\
a_1^* &= \frac{I(C_{\alpha,f} + C_{\alpha,r}) + mv_x(C_{\alpha,f}l_f^2 + C_{\alpha,r}l_r^2 - C_{\alpha,f}l_f\sigma_r + C_{\alpha,r}l_r\sigma_f)}{Im\sigma_f\sigma_r} \\
a_2^* &= \frac{I(mv_x^2 + C_{\alpha,f}\sigma_r + C_{\alpha,r}\sigma_f) + m(C_{\alpha,f}l_f^2\sigma_r + C_{\alpha,r}l_r^2\sigma_f)}{Im\sigma_f\sigma_r} \\
a_3^* &= \frac{(\sigma_f + \sigma_r)v_x}{\sigma_f\sigma_r} \\
a_4^* &= 1
\end{aligned}$$

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